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### Dual-Channel, 0V to 76V Current-Sense Amplifier

# **MAX40201**

### **General Description**

The MAX40201 dual-channel, high-side, current-sense amplifier has V<sub>OS</sub> of less than 12µV (max) and gain error of less than 0.1% (max). The outputs of the MAX40201 are voltage outputs capable of swinging to a few tens of millivolts of the V<sub>DD</sub> rail and ground.

The MAX40201 features an input common-mode voltage range from 0V to 76V with 80kHz of small-signal bandwidth, which makes it ideal for interfacing with a SAR ADC for multichannel multiplexed data acquisition systems.

The MAX40201 operates over the -40°C to +125°C temperature range. The MAX40201 is offered in 8-bump wafer-level package (WLP) and 8-pin  $\mu$ MAX<sup>®</sup> package.

### **Applications**

- Base Stations and Communication Equipment
- Power Management Systems
- Server Backplanes
- Industrial Control and Automation

### **Benefits and Features**

- 0V to 76V Input Common Mode
- Low 12µV (max) Input Offset Voltage
- Low 0.1% (max) Gain Error
- Gain Options
  - G = 25V/V (MAX40201T)
  - G = 50V/V (MAX40201F)
  - G = 100V/V (MAX40201H)
  - G = 200V/V (MAX40201W)
- 1.3mm x 2mm 8-Bump WLP and 8-Pin µMAX Packages

Ordering Information appears at end of data sheet.

 $\mu$ MAX is a registered trademark of Maxim Integrated Products, Inc.

# **Typical Operating Circuit**



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# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Absolute Maximum Ratings**

V <sub>DD</sub> to GND	0.3V to +6.0V
RS+, RS- to GND	0.3V to +80V
RS+ to RS-	±30V
Continuous Input Current (Any Pin)	±20mA
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
WLP (derate 13.4mW/°C above +70°C)	1072mW
µMAX (derate 4.8mW/°C above +70°C)	387.8mW

Operating Temperature Range40°C to +125°C	
Junction Temperature+150°C	
Storage Temperature Range65°C to +150°C	
Lead Temperature (soldering, 10s)(µMAX only)+300°C	
Soldering Temperature (reflow)+260°C	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### Package Thermal Characteristics (Note 1)

#### WLP

Junction-to-Ambient Thermal Resistance  $(\theta_{JA})$  ......74.7°C/W

μMAX

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to <a href="http://www.maximintegrated.com/thermal-tutorial">www.maximintegrated.com/thermal-tutorial</a>.

### **Electrical Characteristics**

 $(V_{RS+} = V_{RS-} = +50V, V_{DD} = +3.3V, V_{SENSE} = V_{RS+} - V_{RS-} = 1mV, T_A = -40^{\circ}C$  to +125°C, unless otherwise noted. Typical values are at T<sub>A</sub> =+25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
DC CHARACTERISTICS				·			
Supply Voltage	V <sub>DD</sub>	Guarante	ed by PSRR	2.7		5.5	V
Supply Current (Both Channels)	IDD	No loads	No loads		1.3	2.1	mA
Power-Supply Rejection Ratio	PSRR	2.7V ≤ V <sub>C</sub>	$2.7V \le V_{DD} \le 5.5V$				dB
Input Common-Mode Voltage Range	V <sub>CM</sub>	Guarante	Guaranteed by CMRR			76	V
Input Bias Current at V <sub>RS+</sub> and V <sub>RS-</sub>	I <sub>RS+</sub> , I <sub>RS-</sub>					0.2	μΑ
Input Offset Current	I <sub>RS+</sub> - I <sub>RS-</sub>					0.2	μA
Input Leakage Current	I <sub>RS+</sub> , I <sub>RS-</sub>	$V_{DD} = 0V$	V <sub>DD</sub> = 0V, V <sub>RS+</sub> = 76V			0.2	μA
Common-Mode Rejection Ratio	CMRR	0V < V <sub>RS</sub>	0V < V <sub>RS+</sub> < 76V				dB
	V <sub>OS</sub>	μΜΑΧ	$T_A = +25^{\circ}C, 0 < V_{CM} < 76V$		0.5	±12	- μV
Input Offset Voltage			$\begin{array}{l} -40^{\circ}\mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq +125^{\circ}\mathrm{C}, \\ 0 < \mathrm{V}_{\mathrm{CM}} < 76\mathrm{V} \end{array}$			±25	
			T <sub>A</sub> = +25°C, 0 < V <sub>CM</sub> < 76V			±20	
		WLP	$\begin{array}{l} -40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq +125^{\circ}\text{C}, \\ 0 < \text{V}_{\text{CM}} < 76\text{V} \end{array}$			±33	
Input Offset Voltage Drift	TCV <sub>OS</sub>					130	nV/°C

# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Electrical Characteristics (continued)**

 $(V_{RS+} = V_{RS-} = +50V, V_{DD} = +3.3V, V_{SENSE} = V_{RS+} - V_{RS-} = 1mV, T_A = -40^{\circ}C$  to +125°C, unless otherwise noted. Typical values are at T<sub>A</sub> =+25°C.) (Note 2)

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
Input Sense Voltage	V <sub>SENSE</sub>	MAX4020	1T (G = 25V/V)		100		
		MAX40201F (G = 50V/V)		50			mV
		MAX402015H (G = 100V/V)		25			
		MAX40201W (G = 200V/V)			12.5		
		Full-scale V <sub>SENSE</sub> = 100mV			25		
		Full-scale	V <sub>SENSE</sub> = 50mV	50			
Gain (Note 3)	G	Full-scale	V <sub>SENSE</sub> = 25mV		100		V/V
		Full-scale	V <sub>SENSE</sub> = 12.5mV		200		
			T <sub>A</sub> = +25°C			0.1	- %
0 · F	05	μΜΑΧ	$-40^{\circ}C \le T_A \le +125^{\circ}C$			0.25	
Gain Error	GE	WLP	T <sub>A</sub> = +25°C			0.15	
			-40°C ≤ T <sub>A</sub> ≤ +125°C			0.3	
Output Resistance	R <sub>OUT</sub>	V <sub>OUT</sub> = V	DD /2, I <sub>OUT</sub> = ±500µA		0.1		Ω
	V <sub>OL</sub>	Sink 500µA				15	
Output Low Voltage		No load				4	mV
Output High Voltage	V <sub>OH</sub>	Source 500µA		V <sub>DD</sub> - 0.015			V
AC CHARACTERISTICS		•					
Signal Bandwidth	BW -3dB	All gain co	onfigurations V <sub>SENSE</sub> > 5mV		80		kHz
AC Power-Supply Rejec- tion Ratio	AC PSRR	f = 200kHz			40		dB
AC CMRR	AC CMRR	f = 200kHz, 100mV sine wave			47		dB
	C <sub>LOAD</sub>	With $250\Omega$ isolation resistor			20		nF
Capacitive Load Stability		Without any isolation resistor			200		pF
Input Voltage-Noise Density	e <sub>n</sub>	f = 1kHz			95		nV/√Hz
Power-Up Time (Note 4)					500		μs
Settling Time (Settling to 0.1%)		$V_{SENSE}$ steps from 20% FS to 80% FS ( $t_R = t_F = 5\mu$ s), $V_{CM} = 24V$ , $C_L = 20pF$			20		μs

Note 2: All devices are 100% production tested at T<sub>A</sub> = +25°C. All temperature limits are guaranteed by design.
 Note 3: Gain and offset voltage are calculated based on two point measurements: V<sub>SENSE1</sub> and V<sub>SENSE2</sub>. V<sub>SENSE1</sub> = 20% x Full Scale V<sub>SENSE</sub>. V<sub>SENSE2</sub> = 80% x Full Scale V<sub>SENSE</sub>.

Note 4: Output is high-Z during power-up.

# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Typical Operating Characteristics**

 $(V_{DD} = 3.3V, V_{RS} + = V_{RS} - = 50V, V_{SENSE} = V_{RS} + V_{RS} - = 1mV, T_A = +25^{\circ}C$ , unless otherwise noted.





# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3.3V, V_{RS\_+} = V_{RS\_-} = 50V, V_{SENSE\_} = V_{RS\_+} - V_{RS\_-} = 1mV, T_A = +25^{\circ}C, unless otherwise noted.$ 













# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3.3V, V_{RS_+} = V_{RS_-} = 50V, V_{SENSE_} = V_{RS_+} - V_{RS_-} = 1mV, T_A = +25^{\circ}C$ , unless otherwise noted.









# Dual-Channel, 0V to 76V Current-Sense Amplifier

### **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3.3V, V_{RS\_+} = V_{RS\_-} = 50V, V_{SENSE\_} = V_{RS\_+} - V_{RS\_-} = 1mV, T_A = +25^{\circ}C, unless otherwise noted.$ 





#### www.analog.com

# Dual-Channel, 0V to 76V Current-Sense Amplifier

# **Pin Configuration**



# **Pin Description**

F	PIN	NAME	FUNCTION
WLP	μΜΑΧ	NAME	FUNCTION
A1	1	RS1+	Channel 1 External Resistor Power-Side Connection
A2	2	RS1-	Channel 1 External Resistor Load-Side Connection
A3	3	RS2+	Channel 2 External Resistor Power-Side Connection
A4	4	RS2-	Channel 2 External Resistor Load-Side Connection
B1	8	V <sub>DD</sub>	Supply Voltage
B2	7	OUT1	Output Channel 1
B3	6	OUT2	Output Channel 2
B4	5	GND	Ground



### Simplified Functional Diagram

### **Detailed Description**

The MAX40201 high-side, current-sense amplifier features a 0V to 76V input common-mode range that is independent of supply voltage. This feature allows the monitoring of current out of a battery as low as 2.7V and enables high-side current sensing at voltages greater than the supply voltage ( $V_{DD}$ ). The MAX40201 monitors current through a current-sense resistor and amplifies the voltage across the resistor.

High-side current monitoring does not interfere with the ground path of the load being measured, making the MAX40201 particularly useful in a wide range of high-voltage systems.

# **Applications Information**

### **Recommended Component Values**

Ideally, the maximum load current develops the full-scale sense voltage across the current-sense resistor. Choose the gain needed to yield the maximum output voltage required for the application:

### $V_{OUT} = V_{SENSE} \times A_V$

where V<sub>SENSE</sub> is the full-scale sense voltage, 100mV for gain of 25V/V, 50mV for gain of 50V/V, 25mV for gain of 100V/V, 12.5mV for gain of 200V/V, and  $A_V$  is the gain of the device.

In applications monitoring a high current, ensure that  $R_{SENSE}$  is able to dissipate its own I<sup>2</sup>R loss. If the resistor's power dissipation exceeds the nominal value, its value may drift or it may fail altogether. The MAX40201 senses a wide variety of currents with different sense-resistor values.

### **Choosing the Sense Resistor**

Choose R<sub>SENSE</sub> based on the following criteria:

**Voltage Loss:** A high  $R_{SENSE}$  value causes the powersource voltage to degrade through IR loss. For minimal voltage loss, use the lowest  $R_{SENSE}$  value.

**Accuracy:** A high  $R_{SENSE}$  value allows lower currents measured more accurately. This is due to offsets becoming less significant when the sense voltage is larger. For best performance, select  $R_{SENSE}$  to provide approximately 100mV (gain of 25V/V), 50mV (gain of 50V/V), or 25mV (gain of 100V/V), 12.5mV (gain of 200V/V) of sense voltage for the full-scale current in each application.

**Efficiency and Power Dissipation:** At high current levels, the  $l^2R$  losses in  $R_{SENSE}$  can be significant. Consider this when choosing the resistor value and its power dissipation (wattage) rating. In addition, the sense resistor's value might drift if it heats up excessively.

# Dual-Channel, 0V to 76V Current-Sense Amplifier

**Inductance:** Keep inductance low if I<sub>SENSE</sub> has a large high-frequency component. Wire-wound resistors have the highest inductance, while metal film is somewhat better. Low-inductance, metal-film resistors are also available. Instead of being spiral wrapped around a core, as in metal-film or wire wound resistors, they are a straight band of metal and are available in values under 1 $\Omega$ .

Take care to eliminate parasitic trace resistance from causing errors in the sense voltage because of the high currents that flow through R<sub>SENSE</sub>. Either use a four terminal current-sense resistor or use Kelvin (force and sense) PCB layout techniques.

### **Power-Supply Bypassing**

Power-supply bypass capacitors are recommended for best performance and should be placed as close as possible to the supply and ground terminals of the device. A typical value for this supply bypass capacitor is 0.1µF (NP0/C0G type) close to the V<sub>DD</sub>/GND bumps. The capacitors should be rated for at least twice the maximum expected applied voltage. Applications with noisy or high-impedance power supplies may require additional decoupling capacitors to reject power-supply noise.

### **Base Station Application Circuit**

An example of a typical application (Figure 1) of this high-voltage, high-precision current-sense amplifier is in base-station systems where there is a need to monitor the current flowing in the power amplifier. Such amplifiers, depending on the technology, can be biased up to 50V or 60V thus requiring a current-sense amplifier like the MAX40201 with high-voltage common mode. The very low input offset voltage of the MAX40201 minimizes the value of the external sense resistor thus resulting in system power-saving.



Figure 1. MAX40201 Used in Base-Station Application

# Dual-Channel, 0V to 76V Current-Sense Amplifier

# **Ordering Information**

PART	GAIN (V/V)	TEMP RANGE	PIN-PACKAGE	TOP MARK
MAX40201TAWA+*	25	-40°C to +125°C	8 WLP	+AAK
MAX40201TAUA+*	25	-40°C to +125°C	8 µMAX	_
MAX40201FAWA+*	50	-40°C to +125°C	8 WLP	+AAL
MAX40201FAUA+	50	-40°C to +125°C	8 µMAX	_
MAX40201HAWA+*	100	-40°C to +125°C	8 WLP	+AAM
MAX40201HAUA+*	100	-40°C to +125°C	8 µMAX	_
MAX40201WAWA+	200	-40°C to +125°C	8 WLP	+AAN
MAX40201WAUA+*	200	-40°C to +125°C	8 µMAX	_

+Denotes a lead(Pb)-free/RoHS-compliant package.

\*Future Product—Contact factory for availablity.

### **Chip Information**

PROCESS: BICMOS

### **Package Information**

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 WLP	W81A1+1	<u>21-100147</u>	Refer to Application Note 1891
8 µMAX	U8+1	<u>21-0036</u>	<u>90-0092</u>

# Dual-Channel, 0V to 76V Current-Sense Amplifier

# **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	5/17	Initial release	_
1	9/17	Adjusted CMRR specification in <i>Electrical Characteristics</i> table and removed future product asterisk from MAX40201FAWA+ in <i>Ordering Information</i> table	2, 11
2	1/18	Updated Ordering Information table	11
3	3/18	Updated Electrical Charactieristics table	2, 3
4	3/18	Updated Ordering Information table	11
5	10/20	Updated package code for 8 µMAX in Package Information	11



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